Microwave Characteristics of a Pseudomorphic High **Electron Mobility Transistor under Electro-Optical Stimulations**

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Abstract—Comprehensive P_{opt} -, V_{GS} -, and V_{DS} -dependent variations of microwave performances $(f_T \text{ and } f_{\max})$ in a PHEMT under electro-optical stimulation are reported for the first time. Under low P_{opt} , microwave characteristics are observed to be predominantly modulated by the photoconductive effect through the transconductance. Under high optical power. however, they are limited by the photovoltaic effect through the gate capacitance and a parasitic MESFET activated parallel to the In_{0.13}Ga_{0.87}As channel PHEMT. Contrary to the dc current-voltage (I-V) characteristics, which are predominantly controlled by the photoconductive effect with a strong nonlinearity due to a parallel conduction, microwave characteristics strongly depend on the photovoltaic effect as well as the photoconductive effect under electro-optical stimulation. An extended small-signal photonic-microwave model is suggested for better description of PHEMT's under electro-optical stimulations.

Index Terms—Cut-off frequency, HEMT, microwave, photonics, optical control.

I. INTRODUCTION

♥ ONSIDERING microwave-photonics as a promising vehicle for high-speed and high-capacity communication systems, a pseudomorphic high electron mobility transistor (PHEMT) is one of the most competing devices for high performance photonic-microwave circuits and systems due to its superior electro-optical performances [1]-[8]. We also expect that variations of photonic-dc/microwave characteristics in PHEMT's under electro-optical co-stimulation can be used as an extended nondestructive characterization technique for physical mechanisms in the failure and degradation of heterostructure field effect transistors [9].

In this work, photonic-microwave characteristics of an n-channel Al_{0.3}Ga_{0.7}As/GaAs/In_{0.13}Ga_{0.87}As PHEMT are reported for possible applications both as a photonic-microwave responsive device and as a novel method for the characterization of traps-related mechanisms in PHEMT's with multiple

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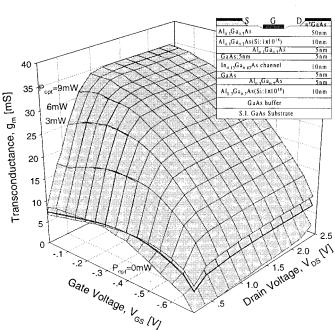
20 15 10 5 2.5 2.0 N 1.5 2.0 N 1.5 V05 Drain Voltage, V05 0 Gate Voltage, V ^{GS} [V] - 3

Fig. 1. Transconductance characteristics of a PHEMT (a V-shaped gate with $L_g = 1 \ \mu \text{m}, L_{\text{gd}} = L_{\text{gs}} = 1 \ \mu \text{m}, \text{and } W = 2 \times 100 \ \mu \text{m}$ as in the inset) under electro-optical stimulations ($P_{opt} = 0, 3, 6, and 9 mW$).

hetero-interface epitaxial structures. The current gain cut-off frequency (f_T) and the maximum frequency of oscillation $(f_{\max(MAG)})$ have been characterized for various electrical operating conditions and optical powers over $P_{\rm opt} = 0 \sim 30$ mW with $\lambda = 0.83 \,\mu$ m. Physical mechanisms involved in the variation of photonic microwave characteristics are analytically explained and an extended photonic-microwave equivalent circuit model is suggested for better description of microwave performances of PHEMT's under electro-optical stimulation.

II. MICROWAVE CHARACTERISTICS OF A PHEMT UNDER **ELECTRO-OPTICAL STIMULATION**

An n-channel Al_{0.3}Ga_{0.7}As/GaAs/In_{0.13}Ga_{0.87}As PHEMT, which has a V-shaped gate with $L_g = 1 \ \mu m, W = 2 \times 100$ μ m, $L_{\rm gs}~=~L_{\rm gd}~=~1~\mu$ m as shown in Fig. 1, was used for an investigation of photonic-microwave characteristics under electro-optical stimulations [10]. Transconductance $(g_m = \partial I_d / \partial V_{\rm gs})$ characteristics are also shown in Fig. 1 for optical powers $P_{opt} = 0 \sim 30$ mW. Electro-optical effects on the drain current (I_D) and the transconductance, are saturated



 $\begin{array}{c} \mbox{TABLE I} \\ \mbox{Highest Values of Current Gain} \\ \mbox{Cut-Off Frequency} \left(f_{T \mid \max} \right), \mbox{Maximum Frequencies of Oscillation} \\ \left(f_{\max({\rm MAG})\mid\max} \mbox{ And } f_{\max(U)\mid\max} \right) \mbox{ Under Electro-Optical} \\ \mbox{Stimulations} \left(P_{\rm opt} = 0, 3, 20, \mbox{ and } V_{\rm DS} \right). \end{array}$

P _{opt}	$f_T _{\max}(V_{GS}, V_{DS})$	$f_{max(MAG)} _{max}(V_{GS}, V_{DS})$	$f_{max(U)} _{max}(V_{GS}, V_{DS})$
0mW	6.62GHz (0.0,2.4V)	16.89GHz (0.0,2.4V)	3.32GHz (-0.7,2.4V)
3mW	9.31GHz (0.0,1.4V)	31.45GHz (0.0,2.4V)	4.40GHz (-0.7,1.4V)
20mW	10.95GHz (0.0,1.4V)	41.79GHz (0.0,2.4V)	4.88GHz (-0.7,1.4V)
30mW	11.02GHz (-0.1,1.4V)	57.15GHz (0.0,2.4V)	4.86GHz (-0.7,1.4V)

predominantly due to a parallel conduction under high optical stimulation [11]. Microwave characteristics are measured as a function of the gate voltage ($V_{\rm GS}$), the drain voltage ($V_{\rm DS}$), and the optical power ($P_{\rm opt}$). $P_{\rm opt}$ -dependent f_T and $f_{\rm max(MAG)}$ have been investigated from scattering parameters measured by an HP-8510B vector network analyzer.

Under electro-optical stimulations, photonic-microwave characteristics are measured and bias-dependent maximum values of f_T and $f_{\max(MAG)}$ are obtained as summarized in Table I. Without optical illumination $(P_{opt} = 0)$, maximum values of f_T and $f_{\max(MAG)}$ were measured to be 6.62 GHz and 16.89 GHz, respectively. As expected, they depend strongly on both V_{GS} and V_{DS} , primarily due to the modulation of transconductance, parasitic capacitances, and parasitic resistances under electro-optical stimulation. Highest values of f_T and $f_{\max(MAG)}$ under $P_{opt} = 30$ mW were improved to $f_T = 11.02$ GHz and $f_{\max(MAG)} = 57.15$ GHz, respectively. Compared with relatively small change from $f_T = 6.62$ to 11.02 GHz, we observed a significant improvement in $f_{\text{max(MAG)}}$ from 16.89 to 57.15 GHz under $P_{\text{opt}} = 30$ mW. Highest values of both f_T and $f_{\max(\mathrm{MAG})}$ were observed at large $V_{\rm DS}$ and zero gate voltage, which has both a high channel carrier density and a high channel electric field with a small gate capacitance. We also observed sharp increase in $f_{\max(MAG)}$ while saturated f_T with increasing optical power for a specific electrical bias V_{GS} and V_{DS} .

Under optical stimulation, predominant characteristic parameters in the variation of both dc and microwave characteristics of PHEMT's include the transconductance, which is modulated mainly by *the photoconductive effect* due to increased channel carrier concentration, and parasitic capacitances, which are controlled predominantly by *the photovoltaic effect*, at a specific electrical bias. Based on a conventional small-signal model of PHEMT's with R_{sd} as a total source-drain resistance $(R_{sd} = R_s + R_d)$, the current gain cut-off frequency is described by [12]

$$f_T = \frac{g_m}{2\pi} \cdot \frac{1}{(C_{\rm gs} + C_{\rm gd}) \cdot [1 + g_{\rm ds}R_{\rm sd}] + g_m C_{\rm gd}R_{\rm sd}} \quad (1)$$

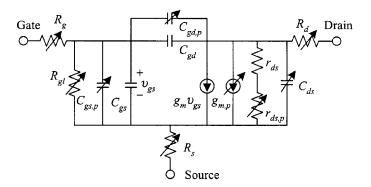


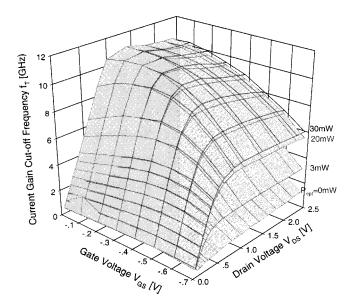
Fig. 2. Extended photonic-microwave equivalent circuit including optically induced transconductance $(g_{m,p})$, gate capacitances $(C_{\text{gs},p}, C_{\text{gd},p})$, output resistance enhancement $r_{\text{ds},p}$, and gate leakage as a parallel resistor (R_{gl}) under electro-optical stimulations.

where the gate capacitance $(C_g = WL\varepsilon_{\text{eff}}/d_{\text{eff}})$ and g_m depend strongly on electro-optical stimulations $(P_{\text{opt}}, V_{\text{GS}} \text{ and } V_{\text{DS}})$ as well as the epitaxial structure of the PHEMT under characterization.

The maximum frequency of oscillation $(f_{\max(MAG)})$ can also be described by (2), shown at the bottom of the page [13], where g_{ds} , $C_{gs}(C_{gd})$, R_{ch} , and $R_g(R_s)$ mean the output conductance, the gate-source(-drain) capacitance, intrinsic channel resistance, and parasitic gate(source) resistance, respectively. As expected, $f_{\max(MAG)}$ depends strongly on parasitic resistances and capacitances as well as g_m via f_T and depends strongly on P_{opt} .

Considering variations of characteristic parameters under optical stimulation, an extended photonic-microwave equivalent circuit model is proposed in Fig. 2. In addition to electrically controlled characteristic model parameters $(g_m, C_{gs}, C_{gd}, r_{ds})$ and parasitic resistances (R_g, R_d, R_s) , we included P_{opt} -dependent model parameters $(g_{m,p}, C_{\mathrm{gs},p}, C_{\mathrm{gd},p}, r_{\mathrm{ds},p})$, and $g_{\rm gl} = 1/R_{\rm gl}$) for better description of photonic-microwave characteristics of PHEMT's. As experimentally described in the small-signal model parameters for P_{opt} = 20 mW in [5], gate capacitances $(C_{\rm gs}, C_{\rm gd})$ and output resistance $r_{\rm ds}$ $(1/r_{\rm ds} = g_{\rm ds} = dI_d/dV_{\rm ds})$, as well as transconductance, increase with P_{opt} while parasitic resistances (R_q, R_s, R_d) decrease due to optically generated excess carriers. This extended photonic-microwave model agrees well with experimental observation in PHEMT's under optical stimulation. Under electro-optical stimulation, the optically induced photoconductivity and the photovoltage $(V_{\text{opt}} = V_{P(0 \text{ mW})} - V_P)$, V_P = pinch-off voltage) are dominant parameters in the modulation of photonic-microwave characteristics. The photoconductive effect, due to optically stimulated excess channel carriers above electrically induced channel carriers controlled by V_{GS} , appears predominantly on the parasitic resistances and transconductance $(g_m = g_{mo} + g_{mp})$ which is saturated at high

$$f_{\max(MAG)} = \frac{f_T}{\sqrt{4g_{ds}\left(R_{ch} + \frac{R_s + R_g}{1 + g_m R_s}\right) + \frac{4}{5} \cdot \frac{C_{gd}}{C_{gs} + C_{gd}} \cdot \left(1 + \frac{5}{2} \frac{C_{gd}}{C_{gs} + C_{gd}}\right) \cdot (1 + g_m R_s)^2}}$$
(2)



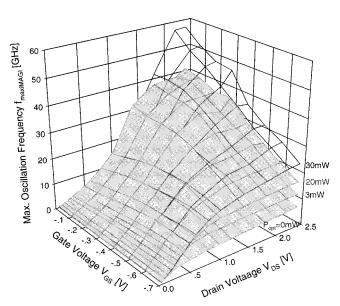


Fig. 3. Current gain cut-off frequencies $(f_T s)$ of a PHEMT as a function of $V_{\rm DS}$ and $V_{\rm GS}$ under electro-optical stimulations ($P_{\rm opt} = 0, 3, 20,$ and 30 mW).

 $P_{\rm opt}$ (>3 mW) as explicitly shown in Fig. 1. $P_{\rm opt}$ -dependent transconductance can be modeled with $g_{\rm mo}$ (transconductance under $P_{\rm opt}=0$) and proportionality constants (a, b, and a') as (3), shown at the bottom of the page, where $\mu_{\rm HEMT}, \mu_{\rm MESFET}$, and $\delta n_{\rm HEMT}$ mean carrier mobility in the InGaAs channel, that in the AlGaAs layer, and saturated carrier density in the InGaAs channel, respectively [11]. $P_{\rm opt,sat}$ and $\delta n_{\rm HEMT,max}$ represent optical power and maximum excess carrier in the InGaAs channel at the onset of parallel conduction.

The photovoltaic effect, which is a result of the accommodation of optical excitation with a depletion of majority carriers at the heterojunction interface, appears mainly on the intrinsic and parasitic capacitances. Considering the optical absorption and excess carrier generation processes in PHEMT, P_{opt} -dependent gate capacitance can be modeled as [10], [11]

$$C_g = C_{\rm go} + C_{\rm gp} \cong C_{\rm go} (1 + \xi' \sqrt{1 + \alpha' P_{\rm opt}}) \qquad (4)$$

with a $P_{\rm opt}$ -dependent gate capacitance $C_{\rm gp}$, a structure-dependent parameter ξ' , and a structure/bias-dependent factor α' [12]. Under low optical stimulation, as a result, the optically induced gate capacitance linearly increases with $P_{\rm opt}$ in addition to the electrically controlled capacitance under $P_{\rm opt} = 0$.

 P_{opt} -dependent f_T is shown in Fig. 3 as a function of P_{opt} , V_{GS} , and V_{DS} . Under low optical power, $f_T - (P_{\text{opt}}, V_{\text{GS}}, V_{\text{DS}})$ curve looks very similar to $g_m - (P_{\text{opt}}, V_{\text{GS}}, V_{\text{DS}})$ curve. One of main causes to the variation of f_T on P_{opt} is an increased transconductance (photoconductive effect) until the photo-induced gate capacitance (photovoltaic effect) plays a dominant role in f_T . Regarding the photoconductive effect on the f_T , the

Fig. 4. Maximum oscillation frequencies $(f_{\max(MAG)})$'s) of a PHEMT as a function of V_{DS} and V_{GS} under electro-optical stimulations ($P_{opt} = 0, 3, 20,$ and 30 mW).

parasitic Al_{0.3}Ga_{0.7}As MESFET, which is known as *a parallel conduction*, is also believed to play a deterministic role in f_T . Under $P_{\rm opt} \leq 3$ mW, the In_{0.13}Ga_{0.87}As channel layer is predominant over the optically induced parasitic MESFET [9]. As a result, the total transconductance is governed by the conduction of In_{0.13}Ga_{0.87}As channel PHEMT. However, the parasitic gate-to-source capacitance $C_{\rm gs}$ doesn't increase so much with $P_{\rm opt} \leq 20$ mW. This agrees well with experimental observation for $f_T - (P_{\rm opt}, V_{\rm GS}, V_{\rm DS})$ because the shape of the $f_T - (P_{\rm opt}, V_{\rm GS}, V_{\rm DS})$ curves is similar to $g_m - (P_{\rm opt}, V_{\rm GS}, V_{\rm DS})$ curve. Contrary to the $g_m - (P_{\rm opt}, V_{\rm GS}, V_{\rm DS})$, which was nearly saturated over $P_{\rm opt} \geq 3$ mW, $f_T - (P_{\rm opt}, V_{\rm GS}, V_{\rm DS})$ curve increases up to $P_{\rm opt} \geq 20$ mW.

Under high optical stimulation with $P_{opt} > 3$ mW, on the other hand, we observed a saturation of g_m resulting in the saturation of f_T with $P_{
m opt}~\geq~20$ mW. This is partially due to the g_m in the saturation region caused by the parallel conduction via the parasitic MESFET with an increased optical power [10], [11]. Although the saturated transconductance does not increase with an additional optical power, the In_{0.13}Ga_{0.87}As channel layer (with a high transconductance) plays a dominant role. So, f_T still increases with elevated optical power up to $P_{\text{opt}} \cong 20 \text{ mW}$. However, the gate capacitance C_g increases with increasing $P_{\rm opt}$ due to a reduced depletion region with an elevated photovoltage under $P_{\rm opt} \geq 20$ mW, which agrees well with previously reported g_m and f_T characteristics [5]. For a given P_{opt} , maximum values of f_T are obtained at a small negative $V_{\rm GS}$ in the saturation mode, which has a saturated maximum drain current over all optical inputs and a small gate capacitance due to the pinched gate-drain capacitance.

$$g_m = \begin{cases} g_{\rm mo} + a\mu_{\rm HEMT}P_{\rm opt} + b\mu_{\rm MESFET}P_{\rm opt}; & P_{\rm opt} \le P_{\rm opt,sat} \\ g_{\rm mo} + a'\mu_{\rm HEMT}\delta n_{\rm HEMT,max} + b\mu_{\rm MESFET}P_{\rm opt}; & P_{\rm opt} > P_{\rm opt,sat} \end{cases}$$
(3)

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Maximum frequencies of oscillation are shown in Fig. 4 as a function of $P_{\rm opt}, V_{\rm GS}$, and $V_{\rm DS}$. $f_{\rm max(MAG)}$ increases significantly with increasing $P_{\rm opt}$, especially at small negative $V_{\rm GS}$ and large $V_{\rm DS}.~f_{\rm max(MAG)}$ increases almost linearly with $P_{\rm opt}$ for given $V_{\rm DS}$ and $V_{\rm GS}.$ The shape of $f_{\rm max(MAG)}$ – $V_{\rm DS}$ curves looks similar to $g_m - V_{\rm DS}$ curves at near pinch-off condition over all P_{opt} . However, it is quite different from the shape of $g_m - V_{\rm DS}$ curves at small negative $V_{\rm GS}$. $f_{\rm max(MAG)}$ increases monotonically for $P_{\rm opt} \geq 3~{\rm mW}$ even though maximum values of f_T change little over $P_{\rm opt} \geq 20$ mW for a specific $V_{\rm GS}$ and $V_{\rm DS}$. In addition, $f_{\rm max(MAG)} - (V_{\rm GS}, V_{\rm DS})$ characteristics for a specific P_{opt} shows significantly different shape from $(f_T - V_{GS}, V_{DS})$ curves as shown in Fig. 3, which has the same shape as $g_m - (V_{GS}, V_{DS})$ curves predominantly being controlled by g_m . As shown in Fig. 4, $f_{\max(MAG)}$ decreases sharply with negative V_{GS} at high optical stimulation while it is relatively less dependent on $V_{\rm GS}$ under low optical power. $f_{\rm max} - V_{\rm DS}$ curves look very similar to $g_m - V_{\rm DS}$ curve under large negative $V_{\rm GS}$ over all $P_{\rm opt}$. This implies that the dominant limiting factor in $f_{\max(MAG)}$ is g_m under large negative V_{GS} for a given P_{opt} . Under small negative V_{GS} , on the other hand, $f_{\rm max(MAG)}$ sharply increases with increasing $V_{\rm DS}$ for given $V_{\rm GS}$ and P_{opt} . This is believed to be mainly due to reduced total gate capacitance by the decreased $C_{\rm gd}$ with expanded pinch-off region deep into the channel from the drain for given P_{opt} and $V_{\rm GS}$ because f_T is almost saturated at high $V_{\rm DS}$ for all $V_{\rm GS}$ and $P_{\rm opt}$.

III. CONCLUSION

Microwave characteristics n-channel of an Al_{0.3}Ga_{0.87}As/GaAs/In_{0.13}Ga_{0.87}As PHEMT under electro-optical stimulations are reported for possible applications as a photonic-microwave device and as a novel characterization technique for traps and interface states in hetero-interfaces. The current gain cut-off frequency strongly depended on both V_{GS} and V_{DS} . Maximum frequencies of oscillation obtained from the maximum available power gain showed maximum values at fixed V_{GS} and V_{DS} over all optical power range. Under low Popt, microwave characteristics were mainly controlled by the transconductance while they are limited by the the gate capacitance and a parasitic MESFET parallel to the $In_{0.13}Ga_{0.87}As$ channel PHEMT under high optical power. In the extended small-signal photonic-microwave model, optically induced characteristic parameters were included for better description of PHEMT's under electro-optical stimulation.

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